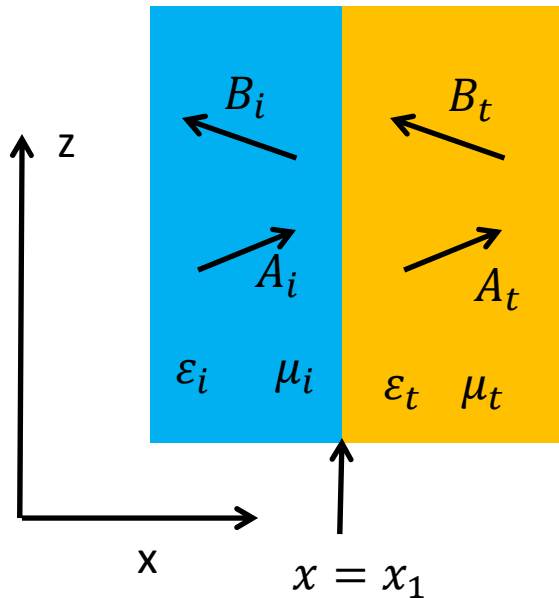


## Case 2: TE, Total Internal Reflection



$$\vec{E}_t = e^{j(\omega t - \beta z)} \{ A_t e^{-jk_{tx}(x-x_1)} + B_t e^{jk_{tx}(x-x_1)} \} \vec{e}_y$$

$$\beta^2 + k_{tx}^2 = \omega^2 \epsilon_t \mu_t$$

With total internal reflection (TIR), we need:  $\beta > \omega \sqrt{\epsilon_t \mu_t}$

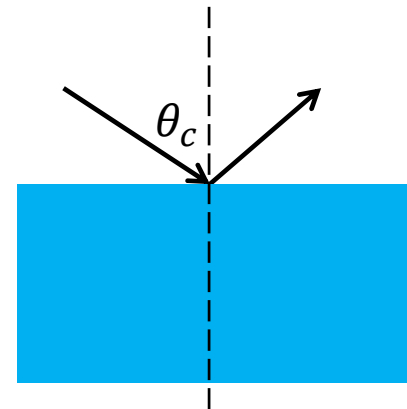
For dielectric material, this generally requires  $\epsilon_i > \epsilon_t$ . The incidence angle also must exceed the critical angle  $\theta_c$  for TIR.

$$n_i = \sqrt{\epsilon_i}$$

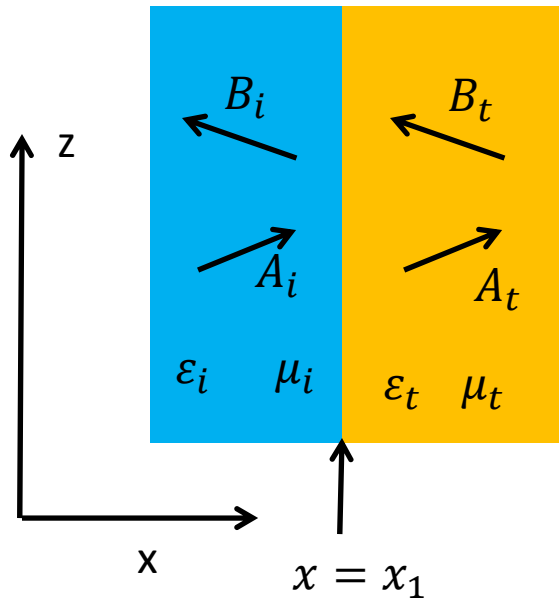
$$n_t = \sqrt{\epsilon_t}$$

$$\mu_i = \mu_t = \mu_0$$

$$\sin \theta_c = \frac{n_t}{n_i}$$



## Case 2: TE, Total Internal Reflection



$$\vec{E}_t = e^{j(\omega t - \beta z)} \{ A_t e^{-jk_{tx}(x-x_1)} + B_t e^{jk_{tx}(x-x_1)} \} \vec{e}_y$$

$$k_{tx} = \sqrt{\omega^2 \epsilon_t \mu_t - \beta^2} = j\kappa_{tx}$$

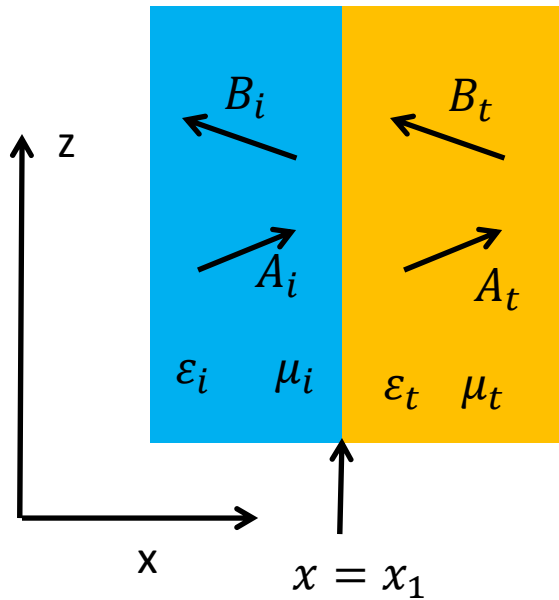
$$\kappa_{tx} = \sqrt{\beta^2 - \omega^2 \epsilon_t \mu_t} > 0$$

$$\vec{E}_t = e^{j(\omega t - \beta z)} \{ A_t e^{\kappa_{tx}(x-x_1)} + B_t e^{-\kappa_{tx}(x-x_1)} \} \vec{e}_y$$

As a result, the boundary condition for the transmission layer is:

$$\begin{bmatrix} A_t \\ B_t \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

## Case 2: TE, TIR



$$D_i^{TE} \begin{bmatrix} A_i \\ B_i \end{bmatrix} = D_t^{TE} \begin{bmatrix} A_t \\ B_t \end{bmatrix}$$

$$D_n^{TE} = \begin{bmatrix} 1 & 1 \\ \frac{k_{nx}}{\mu_n} & -\frac{k_{nx}}{\mu_n} \end{bmatrix}$$

$$\begin{bmatrix} A_i \\ B_i \end{bmatrix} = [D_i^{TE}]^{-1} D_t^{TE} \begin{bmatrix} A_t \\ B_t \end{bmatrix}$$

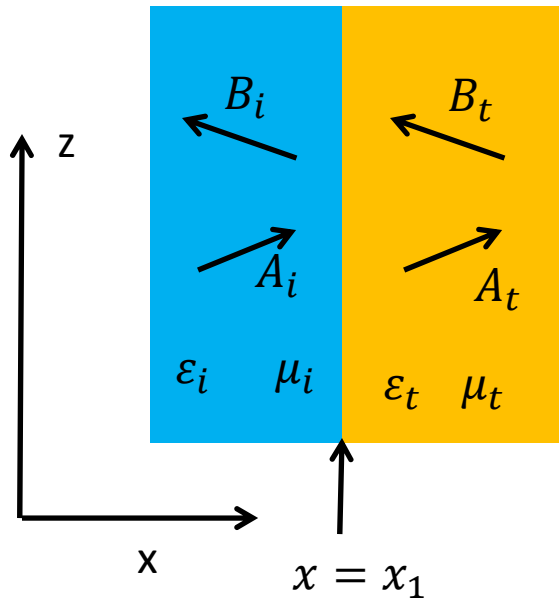
$$k_{tx} = j\kappa_{tx}$$

$$\begin{bmatrix} A_i \\ B_i \end{bmatrix} = \frac{1}{2} \begin{bmatrix} 1 & \frac{\mu_i}{k_{ix}} \\ 1 & -\frac{\mu_i}{k_{ix}} \end{bmatrix} \begin{bmatrix} 1 & 1 \\ \frac{k_{tx}}{\mu_t} & -\frac{k_{tx}}{\mu_t} \end{bmatrix} \begin{bmatrix} A_t \\ B_t \end{bmatrix} = \frac{1}{2} \begin{bmatrix} 1 + \frac{\mu_i k_{tx}}{\mu_t k_{ix}} & 1 - \frac{\mu_i k_{tx}}{\mu_t k_{ix}} \\ 1 - \frac{\mu_i k_{tx}}{\mu_t k_{ix}} & 1 + \frac{\mu_i k_{tx}}{\mu_t k_{ix}} \end{bmatrix} \begin{bmatrix} A_t \\ B_t \end{bmatrix}$$

$$\begin{bmatrix} A_t \\ B_t \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

$$\begin{bmatrix} A_i \\ B_i \end{bmatrix} = \frac{1}{2} \begin{bmatrix} 1 - \frac{\mu_i k_{tx}}{\mu_t k_{ix}} \\ 1 + \frac{\mu_i k_{tx}}{\mu_t k_{ix}} \end{bmatrix}$$

## Case 2: TE, TIR



$$\begin{bmatrix} A_1 \\ B_1 \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

$$\begin{bmatrix} A_i \\ B_i \end{bmatrix} = \frac{1}{2} \begin{bmatrix} 1 - \frac{\mu_i k_{tx}}{\mu_t k_{ix}} \\ 1 + \frac{\mu_i k_{tx}}{\mu_t k_{ix}} \end{bmatrix}$$

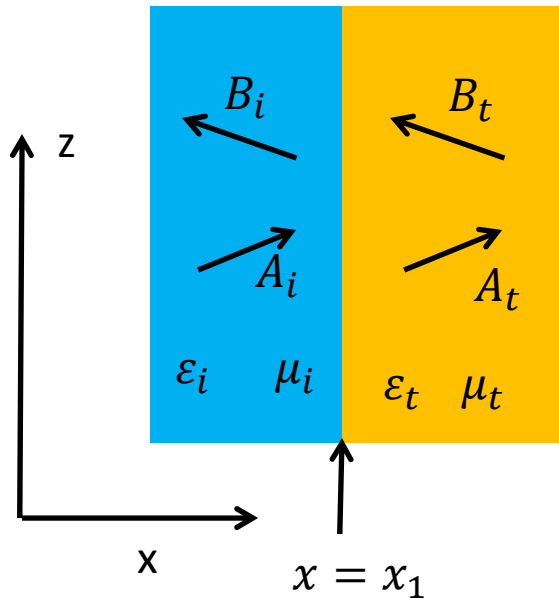
$$k_{ix} = \sqrt{\omega^2 \epsilon_i \mu_i - \beta^2} > 0 \quad k_{tx} = \sqrt{\omega^2 \epsilon_t \mu_t - \beta^2} = j\kappa_{tx}$$

$$r = \frac{B_i}{A_i} = \frac{1 + \frac{\mu_i k_{tx}}{\mu_t k_{ix}}}{1 - \frac{\mu_i k_{tx}}{\mu_t k_{ix}}} = \frac{1 + j \frac{\mu_i \kappa_{tx}}{\mu_t k_{ix}}}{1 - j \frac{\mu_i \kappa_{tx}}{\mu_t k_{ix}}} = e^{j\phi}$$

$$R = \left| \frac{B_i}{A_i} \right|^2 = 1$$

The reflected wave experiences a phase shift but no decrease in amplitude.

## Case 2: TE, TIR



$$\begin{bmatrix} A_1 \\ B_1 \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

$$\vec{E}_t = e^{j(\omega t - \beta z)} B_t e^{-\kappa_{tx}(x-x_1)} \vec{e}_y$$

We can now calculate the Poynting Vector

$$\vec{H}_t = \frac{j}{\omega \mu_t} \nabla \times \vec{E}_t = \frac{j}{\omega \mu_t} \left[ \frac{\partial E_{ty}}{\partial x} \vec{e}_z - \frac{\partial E_{ty}}{\partial z} \vec{e}_x \right]$$

$$H_{tx} = -\frac{j}{\omega \mu_t} \frac{\partial E_{ty}}{\partial z}$$

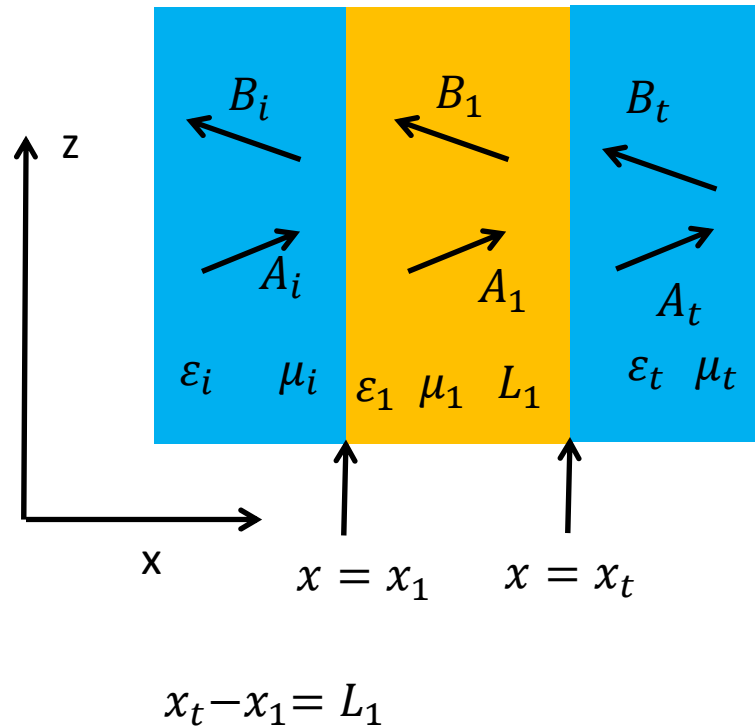
$$H_{tz} = \frac{j}{\omega \mu_t} \frac{\partial E_{ty}}{\partial x}$$

$$H_{tx} = -\frac{\beta}{\omega \mu_t} e^{j(\omega t - \beta z)} B_t e^{-\kappa_{tx}(x-x_1)}$$

$$H_{tz} = -\frac{j \kappa_{tx}}{\omega \mu_t} e^{j(\omega t - \beta z)} B_t e^{-\kappa_{tx}(x-x_1)}$$

$$\vec{S} = \frac{\beta}{2 \omega \mu_t} |B_t|^2 e^{-2 \kappa_{tx}(x-x_1)} \vec{e}_z$$

# Case 3: TE, Dielectric Waveguide



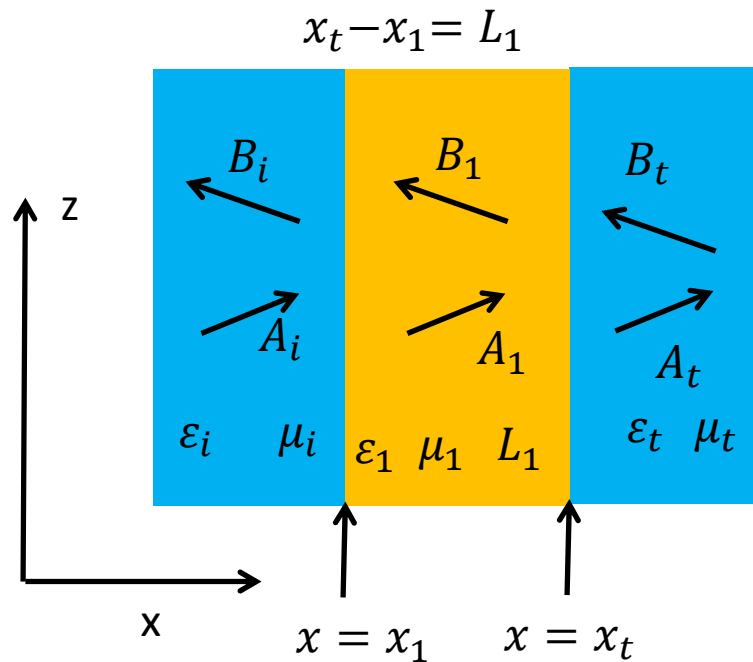
$$D_n^{TE} = \begin{bmatrix} 1 & 1 \\ \frac{k_{nx}}{\mu_n} & -\frac{k_{nx}}{\mu_n} \end{bmatrix}$$

$$P_n^{TE} = \begin{bmatrix} e^{-jk_{nx}L_n} & 0 \\ 0 & e^{jk_{nx}L_n} \end{bmatrix}$$

$$\begin{bmatrix} A_t \\ B_t \end{bmatrix} = T_{TE} \begin{bmatrix} A_i \\ B_i \end{bmatrix}$$

$$T_{TE} = [(D_t^{TE})^{-1} D_1^{TE} P_1^{TE}] [(D_1^{TE})^{-1} D_i^{TE}]$$

# Case 3: TE, Dielectric Waveguide



$$\begin{bmatrix} A_t \\ B_t \end{bmatrix} = \begin{bmatrix} (T_{TE})_{11} & (T_{TE})_{12} \\ (T_{TE})_{21} & (T_{TE})_{22} \end{bmatrix} \begin{bmatrix} A_i \\ B_i \end{bmatrix}$$

$$\vec{E}_t = e^{j(\omega t - \beta z)} \{ A_t e^{-jk_{tx}(x-x_t)} + B_t e^{jk_{tx}(x-x_t)} \} \vec{e}_y$$

$$k_{tx} = \sqrt{\omega^2 \epsilon_t \mu_t - \beta^2} = j\kappa_{tx}$$

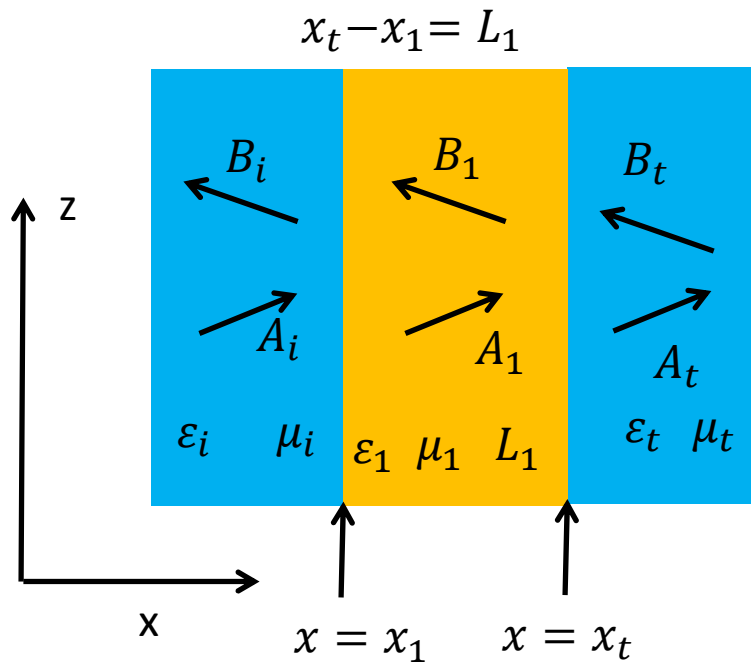
$$\vec{E}_1 = e^{j(\omega t - \beta z)} \{ A_1 e^{-jk_{1x}(x-x_1)} + B_1 e^{jk_{1x}(x-x_1)} \} \vec{e}_y$$

$$k_{1x} = \sqrt{\omega^2 \epsilon_1 \mu_1 - \beta^2} > 0$$

$$\vec{E}_i = e^{j(\omega t - \beta z)} \{ A_i e^{-jk_{ix}(x-x_1)} + B_i e^{jk_{ix}(x-x_1)} \} \vec{e}_y$$

$$k_{ix} = \sqrt{\omega^2 \epsilon_i \mu_i - \beta^2} = j\kappa_{ix}$$

# Case 3: Waveguide Boundary Conditions



$$\begin{bmatrix} A_t \\ B_t \end{bmatrix} = \begin{bmatrix} (T_{TE})_{11} & (T_{TE})_{12} \\ (T_{TE})_{21} & (T_{TE})_{22} \end{bmatrix} \begin{bmatrix} A_i \\ B_i \end{bmatrix}$$

$$\vec{E}_t = e^{j(\omega t - \beta z)} \{ A_t e^{\kappa_{tx}(x - x_t)} + B_t e^{-\kappa_{tx}(x - x_t)} \} \vec{e}_y$$

Boundary Condition:  $A_t = 0$

$$\vec{E}_i = e^{j(\omega t - \beta z)} \{ A_i e^{\kappa_{ix}(x - x_1)} + B_i e^{-\kappa_{ix}(x - x_1)} \} \vec{e}_y$$

Boundary Condition:  $B_i = 0$

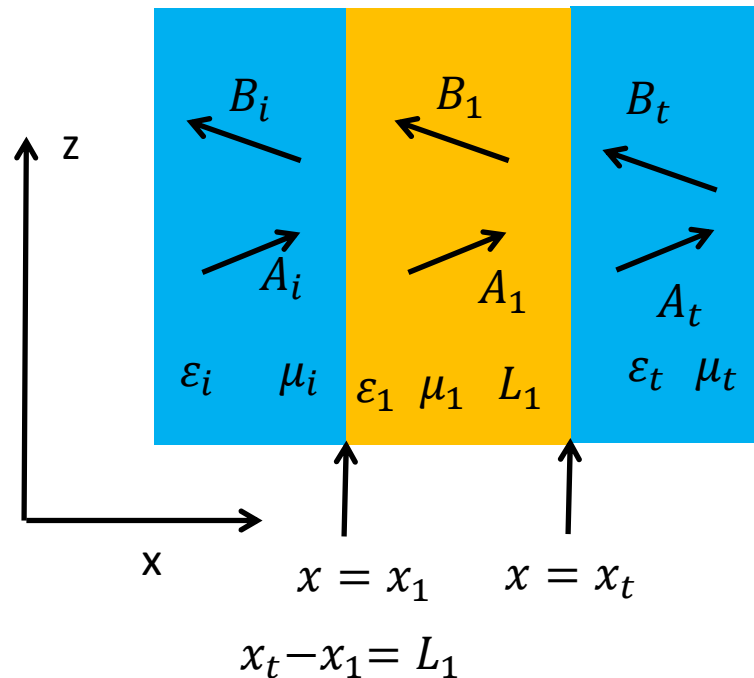
$$\begin{bmatrix} 0 \\ B_t \end{bmatrix} = \begin{bmatrix} (T_{TE})_{11} & (T_{TE})_{12} \\ (T_{TE})_{21} & (T_{TE})_{22} \end{bmatrix} \begin{bmatrix} A_i \\ 0 \end{bmatrix}$$

The equation that determines propagation constant  $\beta$ :

$$(T_{TE})_{11} = 0$$



# Case 3: How to Find Waveguide Modes



$$(T_{TE})_{11} = 0$$

- The equation (highlighted in yellow) is the one that determines waveguide modes.
- If frequency, structural parameters and material properties are given, the transfer matrix will only depend on the value of  $\beta$ .
- Generally speaking, at any given frequency, there are only a discrete number of waveguide modes, each associated with a specific propagation constant  $\beta$ . The number of such modes, however, can be very large depending on waveguide parameters.

$$T_{TE} = [(D_t^{TE})^{-1} D_1^{TE} P_1^{TE}] [(D_1^{TE})^{-1} D_i^{TE}]$$

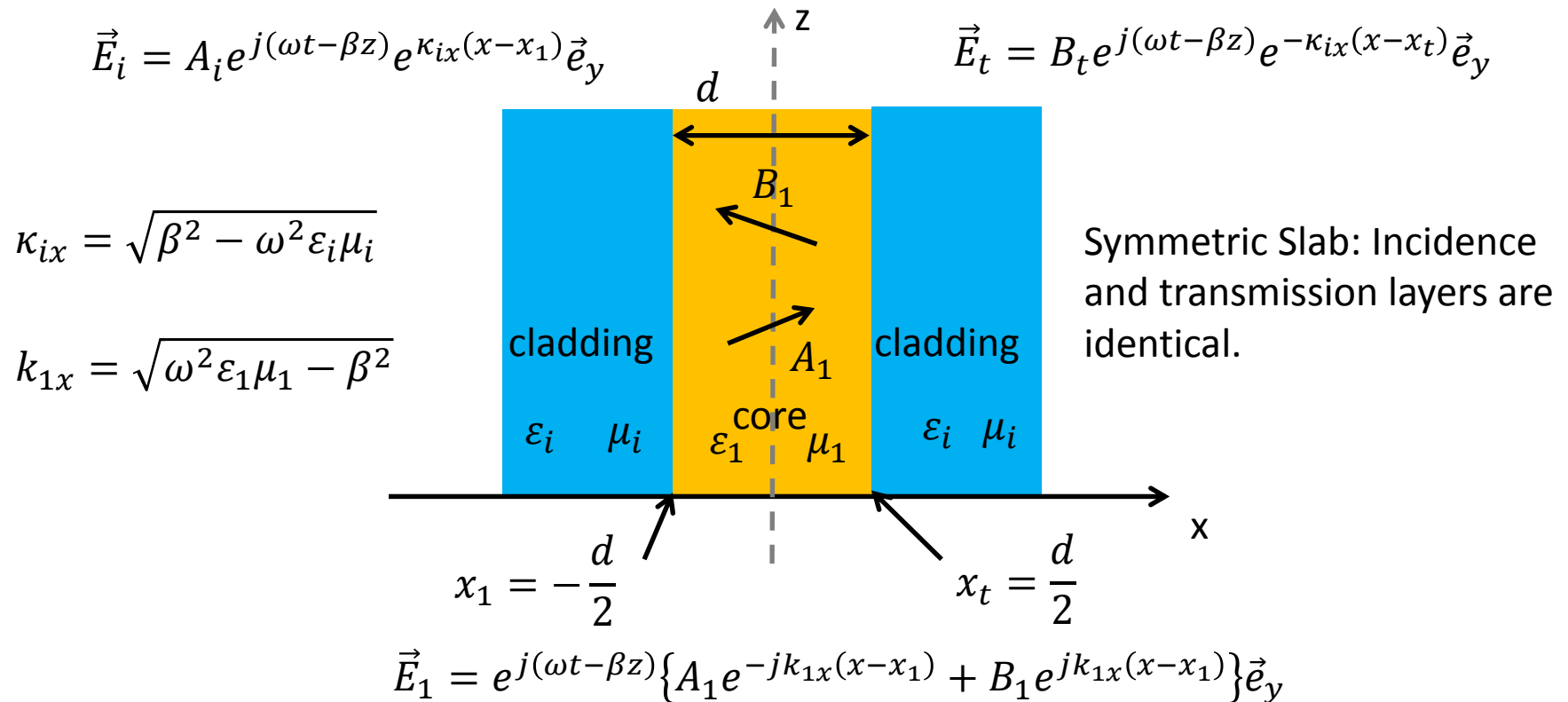
$$D_n^{TE} = \begin{bmatrix} 1 & 1 \\ \frac{k_{nx}}{\mu_n} & -\frac{k_{nx}}{\mu_n} \end{bmatrix} \quad P_n^{TE} = \begin{bmatrix} e^{-jk_{nx}L_n} & 0 \\ 0 & e^{jk_{nx}L_n} \end{bmatrix}$$

$$k_{tx} = \sqrt{\omega^2 \varepsilon_t \mu_t - \beta^2} = j\kappa_{tx}$$

$$k_{1x} = \sqrt{\omega^2 \varepsilon_1 \mu_1 - \beta^2} > 0$$

$$k_{ix} = \sqrt{\omega^2 \varepsilon_i \mu_i - \beta^2} = j\kappa_{ix}$$

# Case 3: TE Modes of a Symmetric Slab

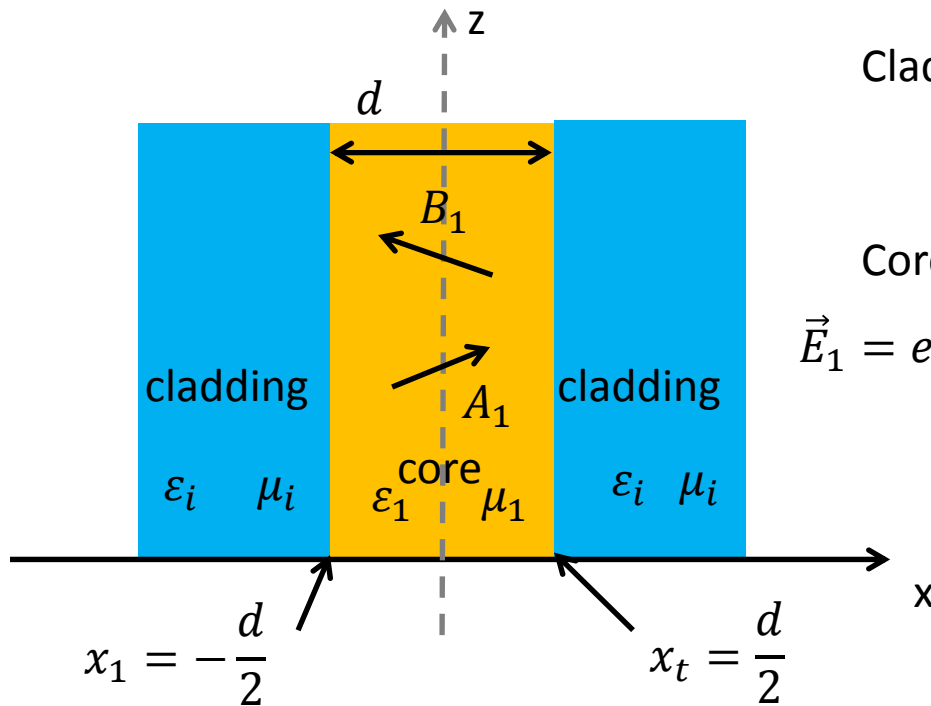


Due to the symmetry of the waveguide geometry, the electric field in the core region can be expressed as either a sin or a cos function.

$$\vec{E}_1 \propto e^{j(\omega t - \beta z)} \cos k_{1x} x \vec{e}_y$$

$$\vec{E}_1 \propto e^{j(\omega t - \beta z)} \sin k_{1x} x \vec{e}_y$$

# Symmetric TE Mode



Cladding:  $\vec{E}_i = A_i e^{j(\omega t - \beta z)} e^{\kappa_{ix}(x-x_1)} \vec{e}_y$

$$B_i = 0$$

Core:

$$\vec{E}_1 = e^{j(\omega t - \beta z)} \{A_1 e^{-jk_{1x}(x-x_1)} + B_1 e^{jk_{1x}(x-x_1)}\} \vec{e}_y$$

In order to achieve the symmetric form:

$$\vec{E}_1 \propto e^{j(\omega t - \beta z)} \cos k_{1x} x \vec{e}_y$$

$$A_1 = e^{jk_{1x}d/2}$$

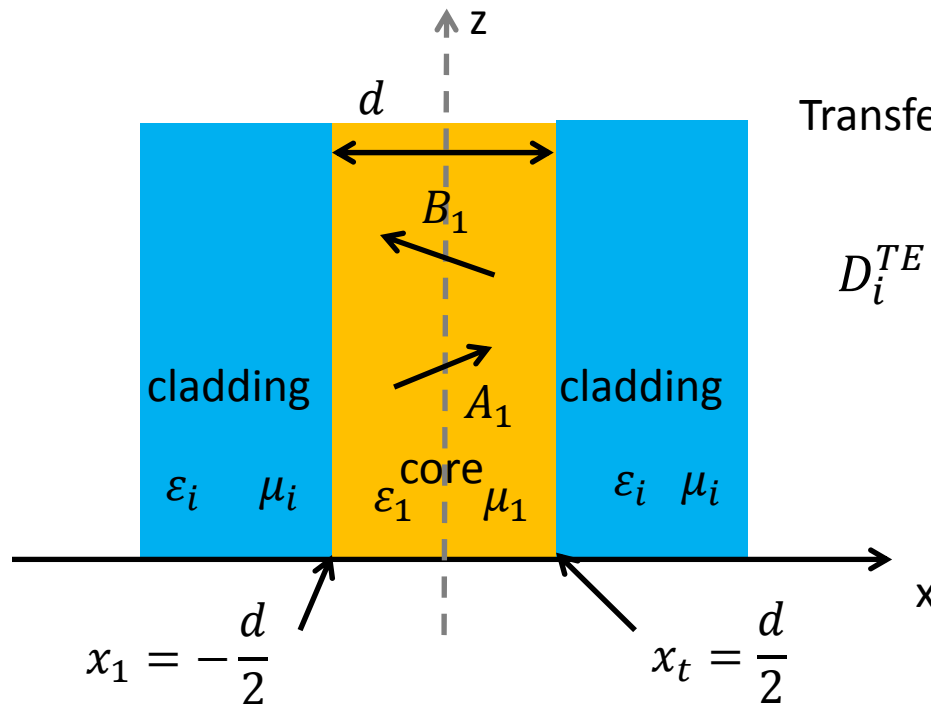
We have:

$$B_1 = e^{-jk_{1x}d/2}$$

$$\kappa_{ix} = \sqrt{\beta^2 - \omega^2 \epsilon_i \mu_i}$$

$$k_{1x} = \sqrt{\omega^2 \epsilon_1 \mu_1 - \beta^2}$$

# Symmetric TE Mode



Transfer matrix between the cladding and core layer:

$$D_i^{TE} \begin{bmatrix} A_i \\ B_i \end{bmatrix} = D_1^{TE} \begin{bmatrix} A_1 \\ B_1 \end{bmatrix}$$

$$D_n^{TE} = \begin{bmatrix} 1 & 1 \\ \frac{k_{nx}}{\mu_n} & -\frac{k_{nx}}{\mu_n} \end{bmatrix}$$

$$\begin{bmatrix} A_i \\ B_i \end{bmatrix} = [D_i^{TE}]^{-1} D_1^{TE} \begin{bmatrix} A_1 \\ B_1 \end{bmatrix}$$

$$\begin{bmatrix} A_i \\ B_i \end{bmatrix} = \frac{1}{2} \begin{bmatrix} 1 & \frac{\mu_i}{j\kappa_{ix}} \\ 1 & -\frac{\mu_i}{j\kappa_{ix}} \end{bmatrix} \begin{bmatrix} 1 & 1 \\ \frac{k_{1x}}{\mu_1} & -\frac{k_{1x}}{\mu_1} \end{bmatrix} \begin{bmatrix} A_1 \\ B_1 \end{bmatrix} = \frac{1}{2} \begin{bmatrix} 1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} & 1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} \\ 1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} & 1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} \end{bmatrix} \begin{bmatrix} A_1 \\ B_1 \end{bmatrix}$$

$$\kappa_{ix} = \sqrt{\beta^2 - \omega^2 \epsilon_i \mu_i} \quad k_{1x} = \sqrt{\omega^2 \epsilon_1 \mu_1 - \beta^2}$$

# Symmetric TE Mode

$$\begin{bmatrix} A_i \\ B_i \end{bmatrix} = \frac{1}{2} \begin{bmatrix} 1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} & 1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} \\ 1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} & 1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} \end{bmatrix} \begin{bmatrix} A_1 \\ B_1 \end{bmatrix}$$

$$B_i = 0$$

$$A_1 = e^{jk_{1x}d/2}$$

$$B_1 = e^{-jk_{1x}d/2}$$

$$\begin{bmatrix} A_i \\ 0 \end{bmatrix} = \frac{1}{2} \begin{bmatrix} 1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} & 1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} \\ 1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} & 1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} \end{bmatrix} \begin{bmatrix} e^{jk_{1x}d/2} \\ e^{-jk_{1x}d/2} \end{bmatrix}$$

$$\kappa_{ix} = \sqrt{\beta^2 - \omega^2 \epsilon_i \mu_i}$$

$$k_{1x} = \sqrt{\omega^2 \epsilon_1 \mu_1 - \beta^2}$$

$$\left(1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}}\right) e^{jk_{1x}d/2} + \left(1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}}\right) e^{-jk_{1x}d/2} = 0$$

$$e^{jk_{1x}d} = -\frac{1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}}}{1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}}} = \frac{\frac{\mu_i k_{1x}}{\mu_1 \kappa_{ix}} + j}{\frac{\mu_i k_{1x}}{\mu_1 \kappa_{ix}} - j}$$

# Symmetric TE Mode

$$e^{jk_{1x}d} = -\frac{1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}}}{1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}}} = \frac{\frac{\mu_i k_{1x}}{\mu_1 \kappa_{ix}} + j}{\frac{\mu_i k_{1x}}{\mu_1 \kappa_{ix}} - j}$$

Set:  $\frac{\cos \frac{\phi}{2}}{\sin \frac{\phi}{2}} = \frac{\frac{\mu_i k_{1x}}{\mu_1 \kappa_{ix}}}{1}$   $e^{jk_{1x}d} = \frac{\cos \frac{\phi}{2} + j \sin \frac{\phi}{2}}{\cos \frac{\phi}{2} - j \sin \frac{\phi}{2}} = e^{j\phi}$   $\phi = k_{1x}d$

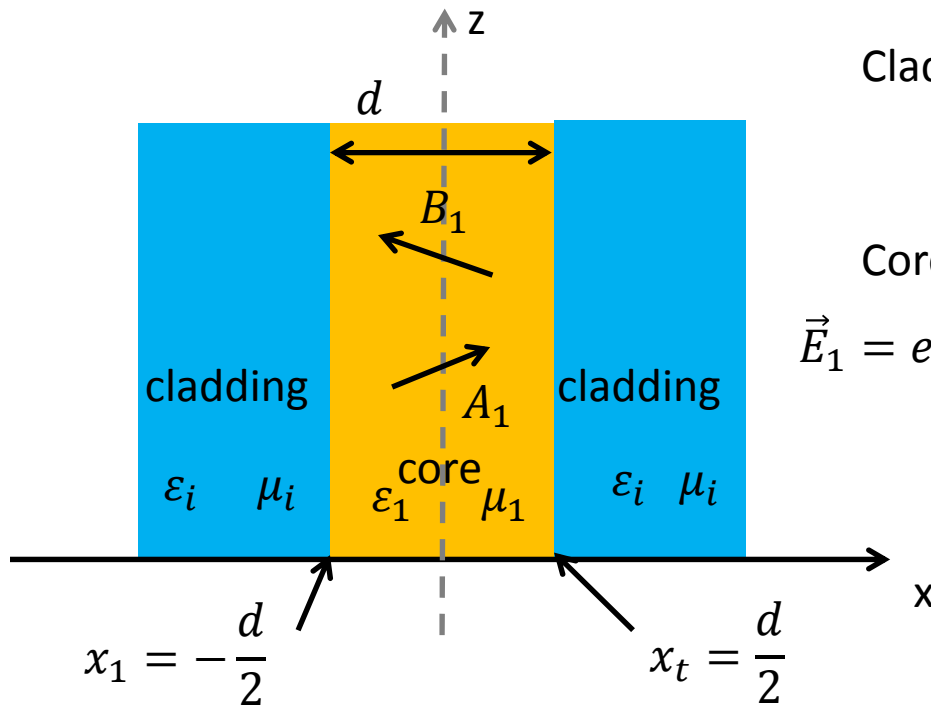
$$\tan \frac{k_{1x}d}{2} = \frac{\mu_1 \kappa_{ix}}{\mu_i k_{1x}}$$

$$\mu_i \frac{k_{1x}d}{2} \tan \frac{k_{1x}d}{2} = \mu_1 \frac{\kappa_{ix}d}{2}$$

$$k_{1x} = \sqrt{\omega^2 \epsilon_1 \mu_1 - \beta^2}$$

$$\kappa_{ix} = \sqrt{\beta^2 - \omega^2 \epsilon_i \mu_i}$$

# Anti-Symmetric TE Mode



Cladding:  $\vec{E}_i = A_i e^{j(\omega t - \beta z)} e^{\kappa_{ix}(x-x_1)} \vec{e}_y$

$$B_i = 0$$

Core:

$$\vec{E}_1 = e^{j(\omega t - \beta z)} \{A_1 e^{-jk_{1x}(x-x_1)} + B_1 e^{jk_{1x}(x-x_1)}\} \vec{e}_y$$

In order to achieve the anti-symmetric form:

$$\vec{E}_1 \propto e^{j(\omega t - \beta z)} \sin k_{1x} x \vec{e}_y$$

$$A_1 = e^{jk_{1x}d/2}$$

We have:

$$B_1 = -e^{-jk_{1x}d/2}$$

$$\kappa_{ix} = \sqrt{\beta^2 - \omega^2 \epsilon_i \mu_i}$$

$$k_{1x} = \sqrt{\omega^2 \epsilon_1 \mu_1 - \beta^2}$$

# Anti-Symmetric TE Mode

$$\begin{bmatrix} A_i \\ B_i \end{bmatrix} = \frac{1}{2} \begin{bmatrix} 1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} & 1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} \\ 1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} & 1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} \end{bmatrix} \begin{bmatrix} A_1 \\ B_1 \end{bmatrix}$$

$$B_i = 0$$

$$A_1 = e^{jk_{1x}d/2}$$

$$B_1 = -e^{-jk_{1x}d/2}$$

$$\begin{bmatrix} A_i \\ 0 \end{bmatrix} = \frac{1}{2} \begin{bmatrix} 1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} & 1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} \\ 1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} & 1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}} \end{bmatrix} \begin{bmatrix} e^{jk_{1x}d/2} \\ -e^{-jk_{1x}d/2} \end{bmatrix}$$

$$\kappa_{ix} = \sqrt{\beta^2 - \omega^2 \epsilon_i \mu_i}$$

$$k_{1x} = \sqrt{\omega^2 \epsilon_1 \mu_1 - \beta^2}$$

$$\left(1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}}\right) e^{jk_{1x}d/2} - \left(1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}}\right) e^{-jk_{1x}d/2} = 0$$

$$e^{jk_{1x}d} = \frac{1 + \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}}}{1 - \frac{\mu_i k_{1x}}{j\mu_1 \kappa_{ix}}} = \frac{\frac{\mu_1 \kappa_{ix}}{\mu_i k_{1x}} - j}{\frac{\mu_1 \kappa_{ix}}{\mu_i k_{1x}} + j}$$



# Anti-Symmetric TE Mode

$$e^{jk_{1x}d} = \frac{\frac{\mu_1 \kappa_{ix}}{\mu_i k_{1x}} - j}{\frac{\mu_1 \kappa_{ix}}{\mu_i k_{1x}} + j}$$

Set:  $\frac{\cos \frac{\phi}{2}}{\sin \frac{\phi}{2}} = \frac{\mu_1 \kappa_{ix}}{\mu_i k_{1x}}$   $e^{jk_{1x}d} = \frac{\cos \frac{\phi}{2} - j \sin \frac{\phi}{2}}{\cos \frac{\phi}{2} + j \sin \frac{\phi}{2}} = e^{-j\phi}$   $\phi = -k_{1x}d$

$$-\cot \frac{k_{1x}d}{2} = \frac{\mu_1 \kappa_{ix}}{\mu_i k_{1x}}$$

$$-\mu_i \frac{k_{1x}d}{2} \cot \frac{k_{1x}d}{2} = \mu_1 \frac{\kappa_{ix}d}{2}$$

$$k_{1x} = \sqrt{\omega^2 \epsilon_1 \mu_1 - \beta^2}$$

$$\kappa_{ix} = \sqrt{\beta^2 - \omega^2 \epsilon_i \mu_i}$$

# Summary of TE Modes

Symmetric:

$$\mu_i \frac{k_{1x}d}{2} \tan \frac{k_{1x}d}{2} = \mu_1 \frac{\kappa_{ix}d}{2}$$

Anti-Symmetric:

$$-\mu_i \frac{k_{1x}d}{2} \cot \frac{k_{1x}d}{2} = \mu_1 \frac{\kappa_{ix}d}{2}$$

d: waveguide thickness.

$\beta$ : propagation constant.

$$k_{1x} = \sqrt{\omega^2 \varepsilon_1 \mu_1 - \beta^2}$$

$$\kappa_{ix} = \sqrt{\beta^2 - \omega^2 \varepsilon_i \mu_i}$$

$$\omega \sqrt{\varepsilon_i \mu_i} < \beta < \omega \sqrt{\varepsilon_1 \mu_1}$$

$$\left( \frac{k_{1x}d}{2} \right)^2 + \left( \frac{\kappa_{ix}d}{2} \right)^2 = \left( \frac{\omega d}{2} \right)^2 (\varepsilon_1 \mu_1 - \varepsilon_i \mu_i)$$

# How to Solve TE Modes

Symmetric:

$$\mu_i x \tan x = \mu_1 y$$

$$x = \frac{k_{1x} d}{2}$$

Anti-Symmetric:

$$-\mu_i x \cot x = \mu_1 y$$

$$y = \frac{\kappa_{ix} d}{2}$$

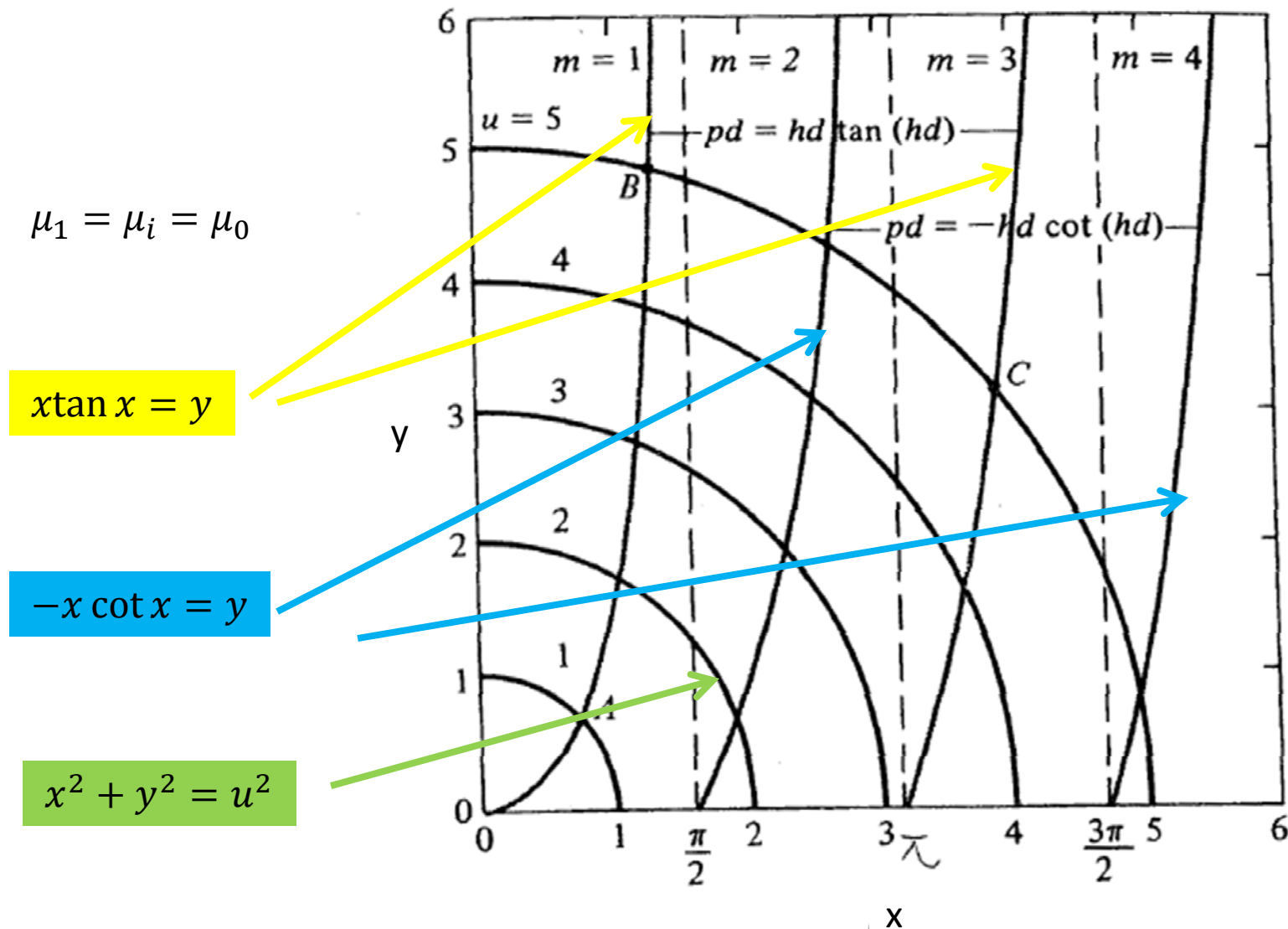
$$x^2 + y^2 = u^2$$

$$u = \frac{\omega d}{2} \sqrt{\epsilon_1 \mu_1 - \epsilon_i \mu_i}$$

For most dielectric materials, we have

$$\mu_1 = \mu_i = \mu_0$$

# Numerical Solution for TE Modes



# How to determine the number of TE Modes

n<sub>2</sub>: cladding index  $n_2 = 1.0$

n<sub>1</sub>: core index  $n_1 = 1.5$

d: waveguide thickness  $d = 1 \mu m$

Wavelength (in air):  $\lambda = 1.55 \mu m$

Non-magnetic materials:  $\mu_1 = \mu_i = \mu_0$

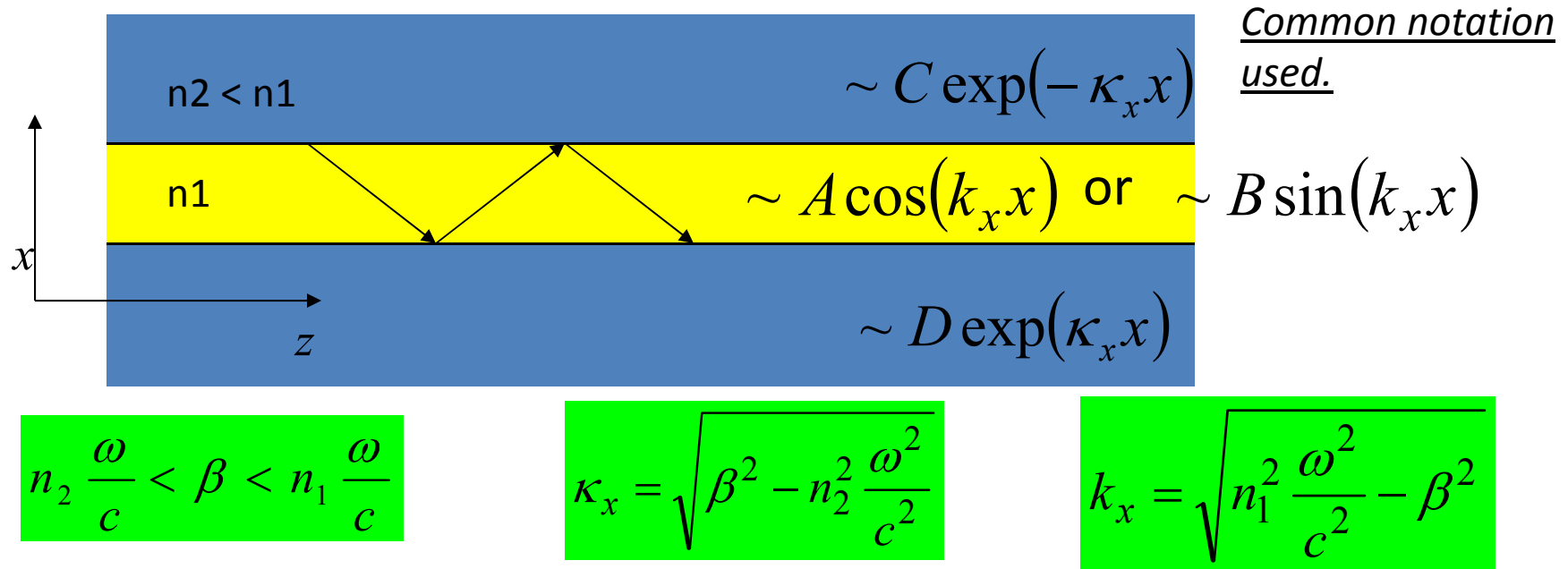
$$\left(\frac{k_{1x}d}{2}\right)^2 + \left(\frac{\kappa_{1x}d}{2}\right)^2 = \left(\frac{\omega d}{2}\right)^2 (\epsilon_1\mu_1 - \epsilon_i\mu_i)$$

$$u = \frac{\omega d}{2} \sqrt{\epsilon_1\mu_1 - \epsilon_i\mu_i}$$

$$u = \sqrt{n_1^2 - n_2^2} \frac{2\pi}{\lambda} \frac{d}{2} = 2.2661$$

TE Modes	Cutoff value for u
TE <sub>1</sub> (symmetric)	$u = 0$
TE <sub>2</sub> (anti-symmetric)	$u = \frac{\pi}{2} = 1.5708$
TE <sub>3</sub> (symmetric)	$u = \pi = 3.1416$
TE <sub>4</sub> (anti-symmetric)	$u = \frac{3}{2}\pi = 4.7124$
TE <sub>5</sub> (symmetric)	$u = 2\pi = 6.2832$

# A Few Comments on Dielectric Waveguide Modes



A few important comments:

1. The value of propagation constant  $\beta$  can only exist between two limits;
2. There exist only a finite number of waveguide modes.
3. The propagation constant  $\beta$  is one of the most important quantity for any waveguide.
4. The physical meaning of  $\kappa_x$  and  $k_x$ . In particular, from their value we can estimate the “incidence angle” and “penetration into the cladding layer”.
5. Note that there a few unknown coefficients in the field expression, and we need to solve certain equations to obtain the propagation constant  $\beta$  and the mode profile.

# Qualitative Analysis of Waveguide Modes

Consider the fundamental TE mode.

If core + cladding index are fixed, at any given wavelength  $\lambda$ ,  $d$  becomes larger and larger.

1. Would  $\beta$  increase or decrease?

- $\beta$  would increase but cannot exceed the limit imposed by:  $\beta < \omega\sqrt{\epsilon_1\mu_1}$

2. Would  $k_x$  &  $\kappa_x$  increase or decrease?

- $k_{1x}d$  would increase but cannot exceed the limit of:  $\frac{k_{1x}d}{2} < \frac{\pi}{2}$
- For a sufficiently large  $d$ ,  $k_{1x}$  would actually decrease.
- $\kappa_{ix}$  would increase but cannot exceed the limit of:  $\kappa_{ix} < \sqrt{\omega^2\epsilon_1\mu_1 - \omega^2\epsilon_i\mu_i}$

3. How would the spatial profile of the waveguide mode change?

The waveguide E field will increasingly look like it will be “pinched off” at the core/cladding boundary. In other words, the E field will increasingly look like a sine function that becomes zero at the core/cladding boundary.